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TM-67-18 SHOCK SPECTRA

by Paul G. Hershall

January 1968





U.S. ARMY MATERIEL COMMAND

HARRY DIAMOND LABORATORIES

WASHINGTON, D.C. 20438

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#### ABSTRACT

Numerical solutions in graphical form are presented for the shock spectra of a one-degree-of-freedom system subjected to some idealized forcing functions. Elementary procedures are outlined for developing the solutions. Analog and digital computer programs are submitted where applicable.

#### 1. INTRODUCTION

In some applications of engineering, knowledge of the maximum displacement of a driven member of a system is desired as a function of a characteristic frequency of the system. This spectrum is called a shock spectrum (ref 1, 2) when transient-type forces, such as in an explosion, are applied to the system.

The system of interest in this report is a one-degree-of-freedom (ODOF) undamped system with a linear restoring force, i.e., a simple mass-spring system. The interest lies in the effect of different driving functions. The study is further limited to a system initially at rest. Ground accelerations and velocities are not considered.

It is not irrelevant whether the maximum displacement occurs in the positive or the negative direction. A structural member generally has different values of Young's modulus in tension and compression. In the situation chosen, the positive maximum is usually larger, but the exceptions will be noted.

Two chief types of forcing functions are considered:

- (1) A half-sine pulse with superimposed Dirac impulses and
- (2) A t<sup>2</sup> exp(-8t) function.

Pulse (1) will have impulses of two types:

- (la) five equispaced, equistrength, positive Dirac impulses with the first at time zero and
- (1b) the preceding with five additional negative impulses symmetrically interspersed (alternating impulses).

A Dirac impulse of strength  $F_1$  will be one for which

$$\int_{0}^{\infty} \mathbf{F}_1 \, \, \delta(\mathbf{t}) \, \mathrm{d}\mathbf{t} = \mathbf{F}_1$$

The scope of this report is limited to presenting the mathematical details used in evaluating particular shock spectra. Their

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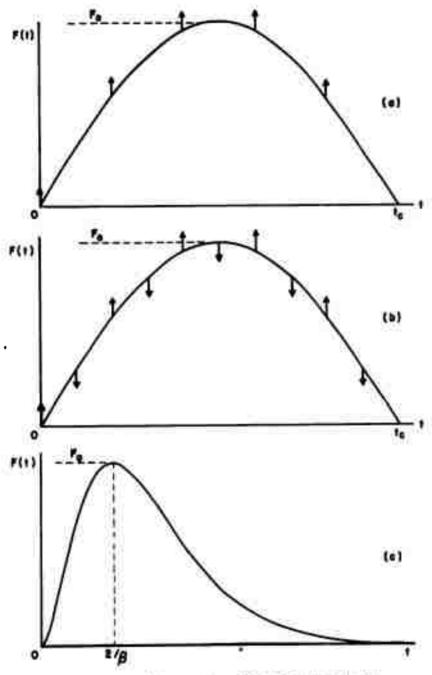


Figure 1. Forcing functions.

relevance will not be pursued here but will be treated in a later report (ref 3). However, one may posit the forcing function having Dirac impulses superimposed on a half-sine pulse as a possible idealization of the high-frequency oscillations often observed on shock pulses (ref 4). Moreover, the penetration resistance of certain projectiles striking steel plates (ref 5) is nearly of the form t exp(-8t).

In section 2, the procedures used to evaluate the shock spectra are described, and general equations are developed. In sections 3 through 6 more detailed information is presented; in section 7 the results for several cases are given.

# 2. GENERAL DEVELOPMENT OF SOLUTION

We are concerned with ODOF systems, i.e., systems described by

$$mx'' + kx = F(t)$$
ditions assumed and (1)

The rest conditions assumed are

$$\mathbf{x}(0) = 0, \quad \mathbf{x}'(0) = 0$$
unctions  $\mathbf{F}(t)$  considered (2)

The forcing functions F(t) considered are illustrated in figure 1.

The deletion of ground accelerations in our ODOF system is not a serious specialization. To clarify this, consider the generalization given in figure 2. Let x be defined by y-z, where y and z are changes in position experienced by the mass m and the ground, respectively, in relation to their rest or reference positions. Then the equation of

$$mx'' + kx = F(t) - mz''$$
ce mz'' is thus equivalent (3)

A ground force mz" is thus equivalent to a force F(t) of opposite sign

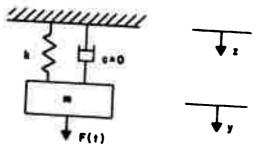


Figure 2. Physical model of generalized ODOF system.

Normalization of the variables in (1) will have two advantages: (1) the equations to be solved are easier to manipulate, having a minimum number of essential variables and parameters; (2) the solutions represent many cases resulting in economy of computation and representation. The first step is to normalize these variables.

One may define

$$\mathbf{F}(\mathbf{t}) = \mathbf{F}_0 f(\mathbf{t}) \tag{4}$$

where  $F_0$  is the peak value of the forcing function excluding impulse components (fig. 1). If we further define

$$v_N^2 = k/m, x_g = F_0/k, u = x/x_g$$
 (5)

equation (1) reduces to

$$\mathbf{u}'' / v_{\mathbf{N}}^2 + \mathbf{u} = \mathbf{f}(\mathbf{t}) \tag{6}$$

Derivatives so far have been with respect to t. In (6),  $m_N$  is the natural frequency of the system,  $x_s$  the static displacement, and u the response factor. We have yet to normalize time; one may, in general, define a unit of time  $t_c$  and let

$$\tau = t/t_{\rm C} \tag{7}$$

Examining figures la and lb, we see that it is most natural to define

$$\gamma_{\mathbf{F}} \mathbf{t_C} = \mathbf{\pi} \tag{8}$$

where  $v_{\mathbf{F}}$  is a frequency characteristic of the driving function. To allow (8) to represent the situation in figure 1c, we may allow

$$\beta = \psi_F/\pi = 1/t_C \tag{9}$$

The characteristic forcing frequency is thus  $\pi\beta$  and the peak value occurs at  $2t_c$ . Furthermore, we normalize the natural frequency as

$$y = w_{N}/w_{F} \tag{10}$$

With these relations, (6) becomes

$$u''(\tau)/\gamma^2\pi^2 + u(\tau) = f(\tau)$$
 (11)

Without ambiguity, derivatives of normalized variables are with respect to  $\tau$ . In addition, some liberty has been taken with the symbols u and f; as functions, they are not the same functions of  $\tau$  that they are of t. The symbols in (11) will, with no real ambiguity, represent the new functions. See appendix D for a table of symbols.

The next step in the analysis is the solution of (11) for u. For the functions presented in figure 1, the solution may be carried out explicitly. There are several known ways of achieving this. The author has chosen the Laplace transform method, but any of the other standard methods are equally quick.

Finally, the maximum displacement is determined for each value of the normalized natural frequency v. For some very simple functions f, the explicit solution u may be differentiated and  $u_{ extbf{M}}$  found as

$$u_{\mathbf{M}} = u(\tau_{\mathbf{M}})$$

where  $\tau_{\mathbf{M}}$  is the solution of  $\mathbf{u}'(\tau) = 0$ . This is not feasible in our cases because of the presence of multiple peaks in u. There appears to be no better recourse than to scan u digitally for a suitable number of values of  $\tau$ , and find the maximax (the global maximum or supremum) or minimax (the supremum in the algebraically negative sense) by comparison. In reference 6, a general program is described to deal with experimental forcing functions. The choice to write an ad hoc program was not easy to make but the ease of finding the response functions analytically, the general clumsiness of large canned programs, and the virtue of added insight with the ad hoc program ruled in

Except for large values of  $\gamma$ , the number of values of  $\tau$  required to scan the peaks are not unreasonable (say~1000) to achieve fourto five-digit precision. The reason for this is that near any peak of u, du/dr is very small, and any error in r is compressed into a much smaller error in u. In view of the use made of the results, it does not seem worthwhile to improve the accuracy by use of local conver-

The preceding remarks apply, in general, to finding  $\mathbf{u}_{\mathbf{M}}$  in the forced-vibration era. It is necessary, also, to find the uM in the free-vibration era (residual spectrum) as for certain values of ; it may be larger than the u<sub>M</sub> in the forced era. In the case of some transient-like functions (cf fig. lc) no free-vibration era exists, strictly speaking. In those cases where it does exist, the spectrum may be simply derived from the forced era response factor u, when damping is absent. Thus, from consideration of constant energy, we must have

$$\frac{1}{2}kx^{2}(t_{c}) + \frac{1}{2}mx^{2}(t_{c}) = \frac{1}{2}kx_{M}^{2}$$
so sen to be to to be some at the second (12)

where t is chosen to be  $t_{C}$  to be sure that all the energy has been added to the system by the external force. The total energy is then equated to the maximum potential energy in the free-vibration era. Normalization of (12) results in the expression

$$u_{\mathbf{M}} = \sqrt{u^{2}(1) + \left(u'(1)/\sqrt{\pi}\right)^{2}}$$
d in obtaining results. (13)

which is used in obtaining results of this study. Sometimes the  $\mathbf{u}_{M}$  for the forced era is not desired. In that case, the alternative method of obtaining the free era spectrum by means of the Fourier spectrum (ref 7) — related by a constant factor to the free spectrum — may

The remaining sections give further mathematical details on the methods and results obtained for the forcing functions considered.

#### HALF-SINE-PULSE-TYPE FORCING FUNCTIONS

As a first specialization of equation (11), let us consider driving functions of the type given in figures la and lb. Since we are dealing with a linear system, we may find the responses to the pure half-sine pulse and the delta function train separately, then add the two proportionately to obtain the total response. (We may not, however, add their separate spectra to find the total spectrum.)

Since we use (13) to obtain uM directly in the free era, the response is needed only for T & 1. The normalized half-sine pulse may therefore be written as

$$f_{HS}(\tau) = \sin \tau \pi \tag{14}$$

and the equation of motion reads

$$u''/v^2\pi^2 + u = \sin \tau \pi \tag{15}$$

By standard methods we obtain from (15) the transfer function
$$U_{HS}(s) = \frac{\gamma^2 \pi^3}{(s^2 + \pi^2)(s^2 + \gamma^2 \pi^2)}$$
(16)

With the aid of a partial-fraction expansion, one obtains the desired inverse transform

$$u_{HS}(\tau) = \frac{\gamma}{1 - \gamma^2} \left( \sin \gamma \tau \pi - \gamma \sin \tau \pi \right) \tag{17}$$

Next, for the finite train of positive delta functions, we have

$$f_{D+}(\tau) = \delta(\tau) + \delta(\tau - \frac{1}{6}) + \cdots + \delta(\tau - \frac{m-1}{6})$$
 (18)

whose transform is

$$F_{D+}(s) = 1 + e^{-s/s} + \cdots + e^{-(m-1)/s/s}$$
 (19)

The function  $f_{D+}$  represents the situation after the  $m^{th}$  impulse and before the  $m+l^{th}$ , where m=1, 2, 3, 4, 5. The train of negative impulses is represented by

$$f_{D_{\tau}}(\tau) = -f_{D_{\tau}}(\tau - \frac{1}{10})$$
 (20)

and has a transform

$$F_{D_{-}}(s) = -e^{-s/1} F_{D_{+}}(s)$$
 (21)

It follows that the transfer functions for these cases -the left side of (15) being still valid --- are

$$U_{D_{\bullet}}(s) = \{ \gamma^{2} \pi^{2} / (s^{2} + \gamma^{2} \pi^{2}) \} F_{D_{\bullet}}(s)$$
 (22)

and

$$U_{D_{-}}(s) = -e^{-8/1} U_{D_{+}}(s)$$
 (23)

whose inverse transforms become

$$u_{DA}(\tau) = -\gamma \pi \{ \sin \gamma \tau \pi + \sin \gamma (\tau - V_E) \pi + \cdots + \sin \gamma (\tau - (m_+ - V_E) \pi \}$$
 (24)

and

$$u_{\mathbf{D}_{-}}(\tau) = -\gamma \pi \{ \sin \gamma (\tau - N_0) \pi + \sin \gamma (\tau - N_0) \pi + \cdots + \sin \gamma (\tau - (2m_{-} - N_0) \pi) \}$$
 (25)

where m<sub>+</sub> and m<sub>-</sub> have the significance of m in (19) and also either  $m_{+} = m_{-}$  or else  $m_{+} = m_{-} + 1$  holds.

The summations over the sine terms in (24) or (25) may be combined by trigonometric identities such as

$$\sin \theta + \sin (\theta - \alpha) + \cdots$$
 m terms =  $\sin \{\theta - (m-1)^{\alpha}/_{3}\} \sin (m\alpha'_{2}) \csc (\alpha'_{3})$ 
is will not be done: the number of  $\sin \theta + \sin \theta = \sin (\theta - (m-1)^{\alpha}/_{3}) \sin (m\alpha'_{2}) \csc (\alpha'_{3})$ 

However, this will not be done; the number of terms in the sum is never greater than five, and the lumped term artificially introduces poles for certain values of  $\gamma$  which require separate handling in a

Thus the final expression for the forced response to the halfsine pulse reads

$$u = u_{HS} + R(u_{D+} + Lu_{D-})$$

$$F_0, \text{ the ratio of the expansion } (26)$$

Here  $R = F_1/F_0$ , the ratio of the strength of a delta function to the amplitude of the pure half-sine pulse; L = 0 for the pulse of figure la, and L = 1 for that of 1b. The programming is discussed in section 5.

The specialization of equation (13) to the response discussed in this section represented by (26) yields for the free-vibracion maximax

$$u_{M} = \sqrt{\{u_{HS}(1) + R[u_{D+}(1) + Lu_{D-}(1)]\}^{2} + \{u'_{HS}(1) + R[u'_{D+}(1) + Lu'_{D-}(1)]\}^{2}}$$
 (13a)  
The substitution of (17), (24), (25)

The substitution of (17), (24), (25), and their derivatives evaluated at  $\tau = 1$  into (13a) was carried out in order to program the residual

# 4. FORCING FUNCTION OF THE TYPE t exp(-x)

The t-dependent function may be written,  $C_0$  being an arbitrary constant, as

$$F(t) = C_0 t^2 e^{-\beta t}$$
is of  $F(t)$  is found to be

The peak value of F(t) is found to be

$$F_0 = 4C_0/\beta^2 e^2 \tag{28}$$

Consequently

$$f(t) = \frac{\beta^{2}e^{2}}{4}t^{2}e^{-\beta t}$$

$$f(\tau) = \frac{e^{2}}{4}\tau^{2}e^{-\tau}$$
(29)

and

$$f(\tau) = \frac{e}{4} \tau^2 e^{-\tau}$$
7) and (9), which together imply that

when we use (7) and (9), which together imply that  $\tau = \beta t_*$ 

The usual liberties have been taken with the symbol f. It is convenient to rescale u and y as

$$u = e^2 w/4, \quad \xi = \gamma \pi \tag{30}$$

Then equation (11) becomes

$$w''/\xi^2 + w = \tau^2 e^{-\tau}$$
 (31)

Using the Laplace transform, one readily obtains
$$W(s) = \frac{2s^2}{(s^2 + 5^2)(s + 1)^3}$$
(32)

One may proceed now by two different paths. The partial fraction expansion may be effected and the inverse transform obtained from the fragments; or the convolution theorem may be applied followed by an integration by parts.

#### 4.1 Partial Fraction Method

Writing

$$\frac{1}{(s^2 + \xi^2)(s+1)^2} = \frac{As + B}{s^2 + \xi^2} + \frac{C}{s+1} + \frac{D}{(s+1)^2} + \frac{E}{(s+1)^2}$$
(33)

one obtains by standard methods (apx A)

$$A = \frac{\xi^2 - 3}{(1 + \xi^2)^3}$$
 (34a)

$$B = \frac{1 - 3\xi^2}{(1 + \xi^2)^3} \tag{34b}$$

$$C = -A (34c)$$

and

$$D = 2/(1+\xi^2)^2 \tag{34d}$$

(34e)  $E = 1/(1+\xi^2)$ 

The solution follows by substitution of these coefficients into the inverse transform obtained from (32) and (33):

$$w(\tau) = 2\varepsilon^{2} \{A\cos \xi \tau + (B/\xi)\sin \xi \tau + (C+D\tau + E^{2\tau^{2}}/2)e^{-\tau}\}$$
 (35)

#### 4.2 Convolution Method

Applying the convolution theorem to equation (32), one obtains the solution to (31) in the form

$$\mathbf{w}(\tau) = \int_{0}^{\tau} \mathbf{v}^{2} e^{-\mathbf{v}} \sin \xi(\tau - \mathbf{v}) d\mathbf{v}$$
 (36)

as may be readily verified by direct substitution. This elegant solution is, however, decidedly inconvenient for computation. As a check on the partial fraction method, one may convert (36) into the form of (35). More details on this are furnished in appendix B. In any case, equation (B-13) developed there is seen to be equivalent to (35).

For reference, the explicit solution is given below as

$$w(\tau) = \frac{2\xi}{(1+\xi^2)^n} \{ (1-3\xi^2) \sin \xi \tau - \xi(3-\xi^2) \cos \xi \tau \} + \xi[(3-\xi^2) + 2(1+\xi^2)\tau + \frac{1}{2}(1+\xi^2)^2\tau^2]e^{-\tau} \}$$
 (37)

## 5. PROGRAM FOR HALF-SINE-PULSE FUNCTIONS

The FORTRAN IV program does the following:

tabled.

- (1) The response factor u is calculated for  $0 \le \tau \le 1$  and
- (2) The values of u are simultaneously compared to determine the maximaxes  $\mathbf{u}_{M}$  in both positive and negative senses.
- (3) These  $\mathbf{u_{M}}^{\prime}\mathbf{s}$  are tabled as a function of  $\gamma$  , the normalized natural frequency.
  - (4) The positive  $u_{M}$ 's are graphed against  $\gamma$ .
  - (5) The residual  $u_{\mathbf{M}}$ 's are tabled and graphed.

For ease of program checkout, options have been incorporated which, among other possibilities, allow the preceding items to be selected individually. These are noted briefly in the comments to the program MAXU in figure 3. These comments show, moreover, that the positive impulse case can be selected by letting Ml be not-unity; that special test values of  $\gamma$  may be read in when M5 = 1 (used with M2 = 1); and that tables for a reduced number of R values may be chosen when KTEST < KK. The particular values of R desired are represented by the KT array.

A selected number of curves may be plotted by using the feature contained in the linking subroutine PLTOPT: the number of plots is read in; for each plot (indexed by K), M values are read into the ICURVE array; these values represent the M values of R chosen to make up a family of curves.

SCALE6 - a subroutine modified from A. Hausner's SCALE2\* - scales the family of curves for each graph. Thus one may plot five curves on one graph, then re-plot three of these, which are scaled independent of the five.

<sup>\*</sup> One of several subroutines documented at HDL for internal use.

```
DIMENSION TAULICOO), TAUPICIOOO), STAUPICIOOO), SGTPICLOOO), UC1000,5)
     1,CTAUPI(1000),UMAX(5),UMIN(5),R(5),KT(5),UR(5),CGP(10),RA(5),
     2 UMINA(5), UMAXA(5)
      DIMENSION G(500). UM(500,5). URESM(500,5)
    1 FORMATILEIS
 11 FORMAT(8F10.5)
                 20X4HR = F7.4,10X8HGAMMA = F0.2,10X7HUMAX = 1PE15.7,
    2 FORMATI
     1 10x7HUMIN - 1PE15.7//10x4(7x1HT12x1HU9x1)
   12 FORMATI (10x4(4x0PF6.3,1PE20.7)))
                12X5HGAMMAS(7X7HUMAX R=F6.41)
   22 FORMATE
   32 FORMAT(/11x0PF6.2,1P5E20.7)
   42 FORMAT(17X1P5E2G.7)
                 30X39HPOSITIVE IMPULSES
                                                  RESIDUAL SPECTRUMIOXAHR . .
   52 FORMATE
   1 F7.4//)
62 FORMAT((15x3(2x5HGAMMAGX5HURMAX4X)))
   72 FORMAT((15×3(0PFO8.1.1PE16.7)))
   82 FORMATIZOX25HFORCED VIBRATION SPECTRUM//)
  92 FORMATIZC X45HTEN ALTERNATING IMPULSES ON A HALF-SINE PULSE//)
102 FORMATIZO X43HFIVE POSITIVE IMPULSES ON A HALF-SINE PULSE//)
                                                  RESIDUAL SPECTRUMIOXAHR . .
  112 FORMATIBOXAZHALTERNATING IMPULSES
  1 F7.4///)
122 FORMAT(17X5(7X7HUMIN R=F6.4))
  132 FORMATILE 1
              MLMBER OF TAU VALUES
      11
              NUMBER OF GAPHA VALUES
       Ú
              NUMBER OF R VALUES
ALTERNATING IMPULSES
WRITE TABLE OF U VS TAU (FORCED ERA)
WRITE TABLE OF UMAX, UMIN VS GAMMA (FORCED ERA)
00000000
      KK
      M1 - 1
      P2 . 1
      H3 - L
                   PLOT UMAX VS GAMMA (FORCED ERA)
      M4 = 1
                   READ GAMMA IN IREAD CARCI
      MS . 1
                   WRITE TABLE OF URESH VS GAMMA (RESIDUAL ERA)
      M6 = 1
      NT = 1 PLOT URESN VS GAMMA (RESIDUAL ERA)
KTEST - NUMBER OF VALUES OF R FOR WHICH TABLES ARE HANTED
      READ(5,1) 11,JJ,KK
      READ(5.1) M1.M2.M3.M4.P5.M6.M7.KTEST
      READ(5,11) DTAU.DG. (R(K),K=1.KK)
      READ(5.1) (KT(K).K=1.KTEST)
      P1 = 3.1415927
EL = M1
       IF(M1.NE.1) EL-C.
      DO 119 KA-1,KTST
       K . KT(KA)
  119 RA(KA) = R(K)
      00 9 1-1-11
       FI . I
       TAU(1) . FI-CTAU
       TAUPI(I) . PIOTAULI)
       STAUPILL) = SIN(TAUPILL))
       (TAUPI(I) = COS(TAUPI(I))
     9 CONTINUE
       1F(M5.EQ.1) GO TO 60
       DO 19 J=1.JJ
```

Figure 3. Program MAXU and subroutine PLTOPT.

```
6(1) . F1.0G
       19 CONTINUE
          GO TO 70
          MARNING - READ STATEMENT CONTROLLEC BY IF STATEMENT
      40 READ(5,11) (G(J),J-1,JJ)
      70 DO 39 J-1,JJ
DO 29 K-1,KK
UMAX(K) = 0.0
         UMENIK) . 0.0
      29 CONTINUE
         DO 89 M=1,10
        FM - M
FGP - FM-C2/10.
        CGP(M) . COS(FGP)
     89 CONTINUE
        DO 49 1-1-11
SGTP1(1) = SIN(G(J) = TAUPI(1))
PAREN = SGTP1(1) - G(J) = STAUPI(1)
IF(1-GT-100) GO TO 15
        BRICK - 0.0
    GO TO 10
15 [F(1.GT.200) GO TO 25
        IF(M1.EQ.1) GO TO 110
  GO TO 10
110 BRICK = SGTPI(1-193)
    GO TO 10
25 [F(1.GT.300) GO TO 35
       IF(M1.EQ.1) GO TO 120
  GO TO 20
120 BRICK - SGTPI(1-100)
       GO TO 20
   35 IF( 1.GT.400) GO TO 45
       IF(M1.EQ.1) GO TO 130
 150 TO 20

130 BRICK = SGTPI(1-103) + SGTPI(1-300)

GO TO 20
  45 IF(1.GT.500) GN TO 55
      IF(M1.EQ.1) GO TO 140
      GO TO 30
 140 BRICK - SGTP1(1-100) + SGTP1(1-300)
      GO TO 30
  55 IF(1.GT.400) GO TO 65
     IF(M1.EQ.1) GO TO 150
     GO TO 33
150 BRICK - SCTP[[1-100] + SCTP[[1-300] + SCTP[[1-500]
 65 IF(1.GT.700) GO TO 75
     IFIM1. EQ. 11 GO TO 162
GO TO 40
160 BRICK = SGTPI(I-100) + SGTPI(I-300) + SGTPI(I-500)
```

mine the standard of the stand

Figure 3. Program MAXU and subroutine PLTOPT.

```
75 [F(].GT.800) GO TO 85
    IF(M1.EQ.1) GO TO 170
170 BRICK - SGTPI(1-100) + SGTPI(1-300) + SGTPI(1-500) + SGTPI(1-700)
    GO TO 40
    GO TO 40
 85 IF(1.GT.900) GO TO 95
    IF(M1.EQ.1) GO TO 189
    GO TO 50
180 BRICK = SGTP1(1-100) + SGTP1(1-300) + SGTP1(1-500) + SGTP1(1-700)
    GO TO 50
 95 IF(M1.EQ.1) GO TU 190
    GO TO 50
190 BRICK + SGTPI(1-100) + SGTPI(1-300) + SGTPI(1-500) + SGTPI(1-700)
   1 + SGTP1(1-990)
    GO TO 50
 10 BRACK - SGTP1(1)
    60 TO 103
 20 BRACK . SGTP1(1) + SGTP1(1-200)
    GO TO 100
 30 BRACK - SGTPI(1) + SGTPI(1-200) + SGTPI(1-400)
 40 BRACK = SGTPI(1) + SGTPI(1-200) + SGTPI(1-400) + SGTPI(1-600)
    GO TO 100
 50 BRACK = SGTPI(1) + SGTPI(1-203) + SGTPI(1-400) + SGTPI(1-690) +
    GO TO 100
   1 SGTP1(1-833)
100 HS . CL.PAREN
    IF(ABS(G(J)-1.3).LT.1.E-5) HS=3.5+(STAUPI(I)-TAUPI(I)+CTAUPI(I))
    DELTP . CZ-BRACK
    DELTH - CZ+BRICK
    DO 59 K-1,KK
    U(I+K) + HS + R(K)+(DELTP - EL+DELTH)
     IF(U(1.K).GE.UMAX(K)) UMAX(K)=U(1,K)
     IFIU(I.K).LE.UMIN(K)) UMIN(K)=U(I.K)
     LM(J.K) . UMAX(K)
 59 CONTINUE
 49 CONTINUE
     UR(K) - C1+(CGP(10)+1.) + C2-R(K)+(CGP(10)+CGP(8)+CGP(6)+CGP(4)+
   1 CGP(2)-EL+(CGP(9)+CGP(7)+CGP(5)+CGP(3)+CGP(1)))
     URESM(J.K) = SQRT(U(11.K)+U(11.K) + UR(K)+UR(K))
  99 CONTINUE
     DO 69 KA-L.KTEST
     K . KT(KA)
     UMAXA(KA) = UMAX(K)
     UMINAIKA) - UMINIK)
     IF(M2.NE.1) GO TO 69
     10 = 0
   7 MRITE(6,132)
     IF(M1.EQ.1) MRITE(6.92)
IF(M1.NE.1) WRITE(6.102)
     WRITE(6,82)
     WRITE(6.2) R(K).G(J).UMAX(K).UMIN(K)
     11 . 1 + IC-200
     12 - 11 + 49
     PRITE(6.12) (TAU(1).U(1.K).TAU(1.50).U(1.50.K).TAU(1.100).
    1 U(1+130.K) . TAH((1+150) .U(1+150.K) . I=11.12)
```

Figure 3. Program MAXU and subroutine PLTOPT.

```
IFIIC.GE.41 GO TO 69
          1C = 1C + 1
GD TO 7
       69 CUNTINUE
          IF(M3.NE-1) GO TO 39
          IF(MOD(J.17).NT.1) GO TO 17
          WRI TE (6 . 132)
          WRITE(5,82)
          IF(MI.EQ.1) WRITE(6.92)
IF(MI.NE.1) WRITE(6.102)
          WRITE(6.22) [RA(K),K=1,KTEST)
         WRITE(6,122) (RA(K),K=1,KTEST)
      17 WRITE(6,32) G(J), (UMAXA(K),K=1,KTEST)
         WRITE(6,42) (UMIVA(K) . K=1.KTEST)
      39 CONTINUE
         IF(M6.NE.1) GO TO 125
         CO 139 KA-1, KTEST
         K . KT(KA)
         ICE . 0
    200 (1 + 1 + 150+ICE
         WRITE (6 , 132)
         IF(M1.EQ.1) WRITE(6,112) R(K)
        IF(M1. NE. 1) WRITE(6,52) R(K)
        WRITE (6,62)
        WRITE(6,72) (G(L), URESM(L,K),G(L+50),URESM(L+50,K),G(L+100),
       1 URESM(L+100.K), L=L1,L2)
        ICE - ICE + 1
IFIICE.LT.31 GO TO 200
   109 CONTINUE
CCC
       WARNING ... PLTOPT HAS READ CARD
  125 IF(M4.EQ.1) CALL PLTOPT(1.G.UM.JJ.KK)
       IF(MT.EQ.1) CALL PLTOPT(2.G.URESM.JJ,KK)
       REWIND 9
       STOP
       END
         SUBRUITINE PLTOPT(L,X,Y,JJ,KK)
DIMENSION X(500),Y(500,5)
         COMMON/DIMENSIN, M.K. ICURVE(5,5) , XAMIN, XAMAX
       1 FCRMAT(1615)
         N = JJ
READ(5.1) NOP! TS
         DC 9 K=1, NOPLTS
         READ(5,1) M
        READ(3,1) (ICURVE(K.J], J=1,M)

IF(L.EQ.1) CALL SCALE6(4,Y,7,10,1,2,0)

IF(L.EQ.2) CALL SCALE6(X,Y,7,10,2,2,0)
      9 CONTINUE
        RETURN
        ENU
```

Figure 3. Program MAXU and subroutine PLTOPT.

The calculations in the bulk of the program are done so as to avoid repetitious calculations of the sine and cosine; arrays such as TAUPI, STAUPI, SGTPI are precalculated, keeping the running time to a reasonable 5 to 8 min on the 7094. A better method, though, uses recurrence formulas (ref 8) to generate all but the first members of the circular functions with multiplications.

When  $\gamma = 1$ , equation (17) has a pole; thus  $C_1$ , after statement 29, becomes zero on the 7094; HS in 100 is thus zero, but the IF statement following it --- allowing for binary roundoff --- calculates the limiting value obtained by L'Hopital's rule.

Many of the variables have names suggestive of their counterparts in the previous sections; hence most of the arithmetic can be followed easily.

#### 6. PROGRAMMING THE texp(-8t) CASE

A simple analysis of the limiting behavior of equation (31) shows that as  $\xi \to \infty$ ,  $w \to \tau^2 e^{-\tau}$ ,  $\tau_{MAX} \to 2$ , and  $w_{MAX} \to 4/e^2 = 0.54134 \cdots$ . When  $\xi \to 0$ ,  $w \to 0$ , and moreover, from (37), when  $\xi \ll 1$ ,

$$w \sim 25(\sin 5\tau - 35\cos 5\tau) + F^2(6 + 4\tau + \tau^2)e^{-\tau}$$
 (38)

which is a sinusoid with a small contribution from the transient term.

It is clear then that for large  $\xi$  we may be certain to find the maximax upon scanning the domain of  $\xi$  between 1 and 3; likewise, for small  $\xi$ , we may expect the first peak to be the maximax. The situation for intermediate values of  $\xi$  is not clarified by a brief additional analysis: e.g., the transient term has no relative maximum at the level of approximation given in equation (38). Practically, this question is readily solved by the analog computer. The results of this study (details are furnished in section 6.1) show that for  $\xi < 3$ , the first-peak criterion is valid. Also the criterion for  $\xi > 3$  is that  $1 < \tau_{MAX} < 3$ .

Based on these two criteria, a straightforward digital program was written in FORTRAN to calculate and plot the shock spectrum. A short explanation of the program is given in section 6.2.

#### 6.1 Analog Computer Analysis

An elementary circuit is sufficient for generating  $w(\tau)$ . All the intermediate functions required can be generated by linear equipment (ref 9). The unscaled equations are:

 $G_1 = e^{-\tau}$   $G_1' = -kG_1$   $G_2' = k(G_1 - G_2)$   $G_3' = k(2G_2 - G_3)$   $H_1' = k\xi^2(G_3 - w)$   $w' = kH_1$  $\tau' = k = d\tau/ds$ 

where all derivatives are with respect to s, the machine time. The scaled circuit is given in figure 4, in which  $S_1$  and  $C_W$  are unknown scaling factors that depend on  $\xi$ . A table giving suitable values of these factors is given in appendix C. The constant c determines the arm speed of the recorder and thus the paper abscissa scaling.

Representative response curves for low, intermediate, and high values of F are given in figure 5.

The ease of programming the differential equation leading to these response curves suggests the use of the repetitive operation feature available on the EAI 231R analog computer. (The 131R was used here.) In this PEPOP technique, the differential equation would be solved in fast time, and the parameter f swept in slow time. The dynamic range of f and a fortiori f is unfortunately too great for the analog computer; scaling problems arise. It appears one can do no better than divide the f-axis into sections and scale each section separately. This naturally vitiates much of the attractiveness of the method, which was therefore not pursued.

### 6.2 Digital Computer Program

The FORTRAN IV program to calculate the shock spectrum is presented in figure 6. The main branch statement after DO 29 allows calculation and search for a maximum for  $1 < \tau < 3$  when  $\xi > 3$ , storing the maximaxes in WMAX(J). If  $\xi < 0$ , a semiglobal algorithm is chosen, which, starting from  $\tau = 0$ , sweeps past the first maximum in fairly large steps, reverses two steps, cuts the interval by five, resweeps and repeats until the intervals are roughly equal to those used in the other path of the program, again storing the results in WMAX. The SCALIT subroutine, written by A. Hausner\* for simple x-versus-y plotting, scales TAU and WMAX, and writes them on tape (in low density).

The shock spectra for all the forcing functions are presented and discussed in section 7.

<sup>\*</sup> One of several subroutines documented by HDL for internal use.

#### 7. RESULTS

The chief results of this study are the shock spectra shown in figures 7 through 12. As can be seen, figure 7 illustrates a family of forced-vibration spectra for different values of R corresponding to the forcing function of figure 1a. Figure 8 is the analogous family corresponding to figure 1b. Figures 9 and 10 are the free-vibration spectra for the same two cases. Figure 11 shows a forced- and a free-vibration spectrum on the same graph for R = 0.01. These are the upper curves in figures 7 and 9. Finally, figure 12 is the spectrum for the forcing function of figure 1c. These results have not been observed in the literature.

The important features of figures 7 through 11 are the resonance peaks near  $\gamma = 1.5$ , 10, 20, and 30. If we were to "continue" the half-sine pulse to a continuous monotonic frequency, then y would measure the natural frequency of the system with respect to the former frequency. When there are five positive impulses per halfsine pulse, the system responds to this repetition rate of ten per "cycle" in giving the resonances at 10, 20, and 30 with negligible augmentation of the major peak of the pure half-sine pulse spectrum (R = 0). Additional energy is fed into the system when interspersed, negative impulses are added; the amplitude of the spectrum is thus larger except for suppression at values of  $\gamma$  at even multiples of 10. The suppression is due to the cancellation of the kinetic energy of the system oscillations by the forcing function. This happens when the system has just completed an oscillation cycle and is then "s ruck" by a negative impulse. The velocity at this moment is maximum in the positive direction, but the force is in the negative direction.

In figure 11, we see how the residual spectrum may predominate for certain values of  $\gamma$ , in this case for  $\gamma < 1$ ; one must in general then consider both the forced and the free era.

The tabular data, not given here, show that in no case is the negative maximax larger than the positive for the forced vibrations. (They are equal in the residual era.) The trend of the data, however, strongly suggests that for resonances at larger  $\gamma$ , the negative maximaxes may become larger than the positive. Moreover, some results generated for R = 0.2 with the earlier program showed that for the alternating impulse case, negative maximaxes were larger near and at the resonances for  $\gamma$  = 10 and 30. Thus, although not investigated here, negative maximaxes may be very significant for large  $\gamma$  and large R.

### 8. REFERENCES

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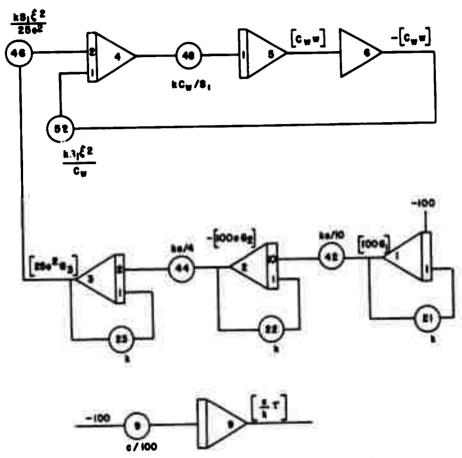


Figure 4. Analog computer circuit for undamped ODOF system forced by  $-^7e^{-7}$ .

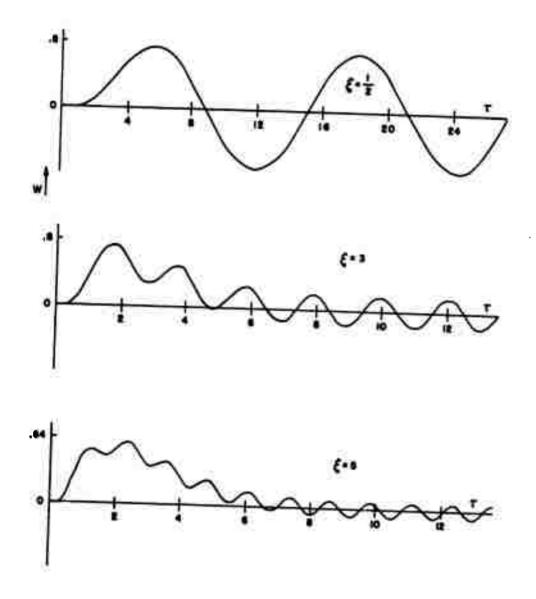


Figure 5. Representative response curves for low, medium and high values.

```
DIMENSION TAU(1000), TAUS(1000), EXMTAU(1000), XI(350), WMAX(350),
      1 A(350).A1(350).A2(350).B1(350).B2(350).B3(350)
     1 FCRMAT(1615)
    11 FCRMAT(0F10.5)
    2 FORMATILHED
    12 FCRMAT((2CX2(7x7HX[13X4HWMAX4X))//)
    22 FCRMAT((2CX2(OPFLO.2,1PE20.7)))
    32 FCRMAT(37X5HII = [4,5X5FJJ = 13)
   42 FCRMAT(37%6HCXI . F5.2,5%7HDTAU . F7.4/37%9HDELTAM . F5.2,5%4HZ .
      1 F7.4///)
       READ(5.1) II.JJ.II.IZ.NX.NY.IDENT.ICOCE
       READ (F.LL) DXI.DTAU, DELTAM, Z
       II - NUMBER OF VALUES COMPUTED OF EACH RESPONSE FUNCTION
       JJ . NUMBER CF POINTS IN THE SPECTALM
       11 . PLUTTING OPTION (PLUT WHEY 11-1)
č
       12 . WHITING CPTION (WRITE TABLE OF RESULTS WHEN 12-1)
00000
       DXI . KI INCREMENT
      DTAU = TAL INCREMENT
DELTAM = INITIAL INCREMENT IN TAM REPRESENTING TAU
DELTAM = UPDATED INCREMENT
       2 = ERROR CRITERIEN IN TAU
      DC LUOP 9 GENERATES ARRAYS BASED ON 1.LT.TAU.LT.3
      DC 9 I=1,11
      FI . 1
      TAU(1) . FI-CTAU + 1.
      TAUS(1) = TAU(1)+TAU(1)
      EXMTAU(I) = EXP(-TAU(I))
    9 CENTINUE
      DC 19 JeliJJ
      WPAXIJ) . 0.0
      FJ . J
      XI(J) = FJ.DXI
      XIS = XI(J) ext(J)
      ONEPKS - 1. + XTS
ONPXSS - CNEPKS-LNEPKS
      DAPKSC . CHPKSS+CHEPKS
      A(J) = 2.4X1(J)/04PXSC
      AL(J) = 1. - ?. . x ! S
      A2(J) = -x1(J) = (2.-x15)
      (L)SA+(L)A- = (L)18
      BZ(J) = 4. .XIS/ONPXSS
      B3(J) = XIS/CNEPXS
  19 CENTINUS
      DC 29 Jal,JJ
      IF(XI(J).LT.3.0) GO TO 100
     DC 39 I=1,II
SIGNIT = SIV(NI(J)+TAU(I))
CCSNIT = CCS(NI(J)+TAU(I))
      W = A(J)+(AL(J)+SINXIT + AZ(J)+CUSXIT) + (D1(J) + BZ(J)+TAU(I) +
     1 P3(J) TAUS([]) CXMTAU([])
      IF( M.GE.MMAX(J)) WMAX(J)=W
  39 CONTINUE
      GO TO 27
 100 TAM + U.U
DELTAN + CELTAM
      WHULDL . C.O
  80 WHULD2 . C.O
```

Figure 6. Program TSEXP.

```
60 TAM = TAM + CELTAN
     MATE(L)IX - MATIX
    W = A(J)+(A)(J)+SIN(XITAM) + A2(J)+COS(XITAM))+ (B1(J) + B2(J)+TAM
1 + B3(J)+TAM+TAM)+EXP(-TAM)
     IFIM.LT. WECLCI) GO TO SO
     MHOLD2 - MHOLC1
     GC TO 60
50 IFIUELTAN.LT.Z) GO TO TO
TAM = TAM - 2. ODELTAN
     WHOLDL = MHOLD2
     DELTAN . CELTAN/S.
     GC TO SO
70 MMAX(J) = WHCLD1
29 CENTINUE
    TF(II.NE.L) GC TC 10
CALL SCALIT(XI,WMAX,JJ,NX,NY,IDENT,ICCDE)
REWIND 9
10 IF112.NE.1) STOP
    JCDUNT - C
    WRITE(6,2)
    WRITE(6,32) II.JJ
WRITE(6,42) CXI.CTAU.DELTAM.Z
30 J1 = 1 + 100 - JC7U4T
J2 = J1 + 49
WRITE(6, 12)
    HRITE(6,22) (XI(J), WMAX(J), XI(J+50), WMAX(J+50), J=J1, J2)
    JCOUNT - JCOUNT + 1
IF(JJ.LE.100+JCOUNT) GO TO 20
    WRITE(6,2)
GC TU 30
20 STOP
    ENC
```

Figure 6. Program TSEXP.

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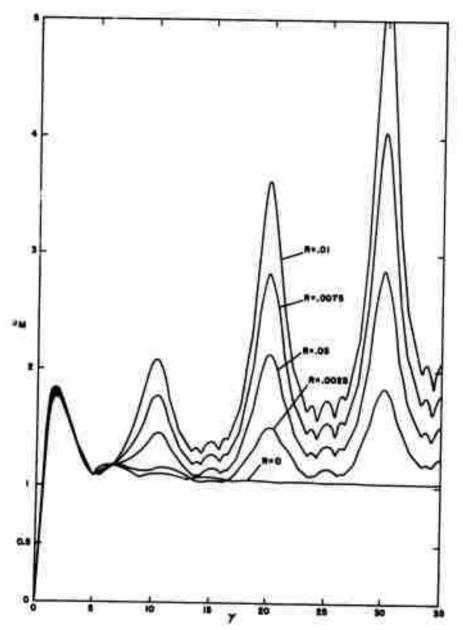


Figure 7. Forced vibration spectra of five positive impulses on a half-sine pulse driving an ODOF system.

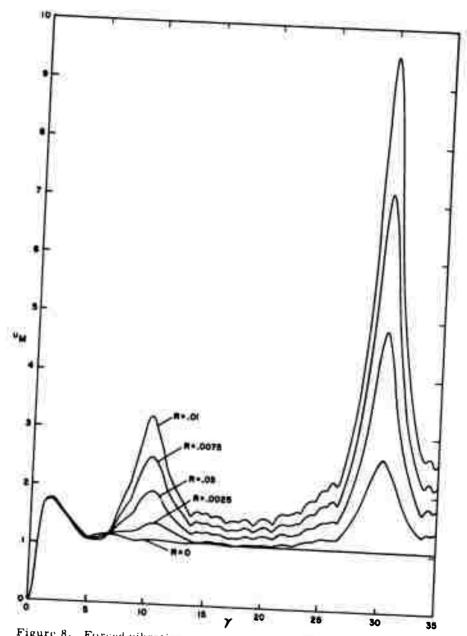


Figure 8. Forced vibration spectra of ten alternating impulses on a half-sine pulse driving an ODOF system.

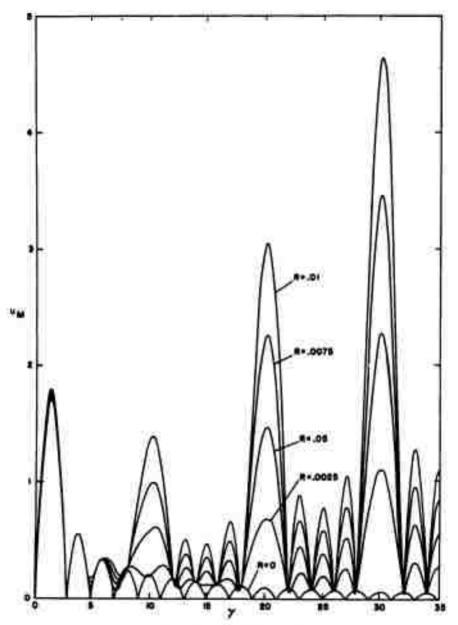


Figure 9. Free vibration spectra of five positive impulses on a half-sine pulse driving an ODOF system.

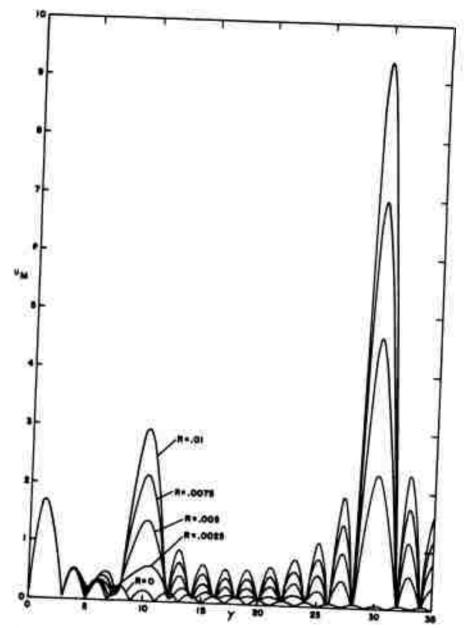


Figure 10. Free vibration spectra of ten alternating impulses on a half-sine pulse driving an ODOF system.

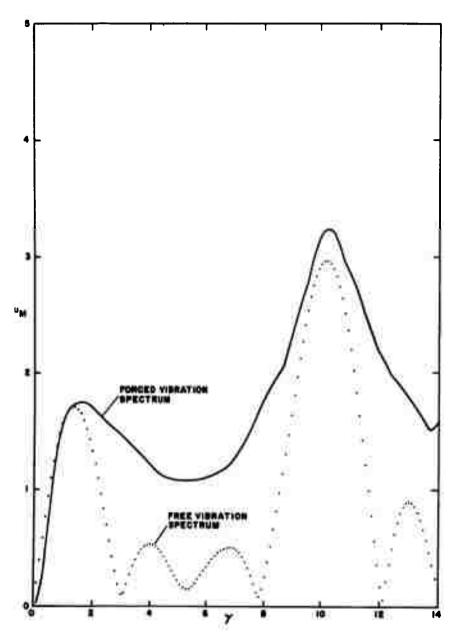


Figure 11. Comparison of forced and free vibration spectra for ten alternating impulses; R = 0.01.

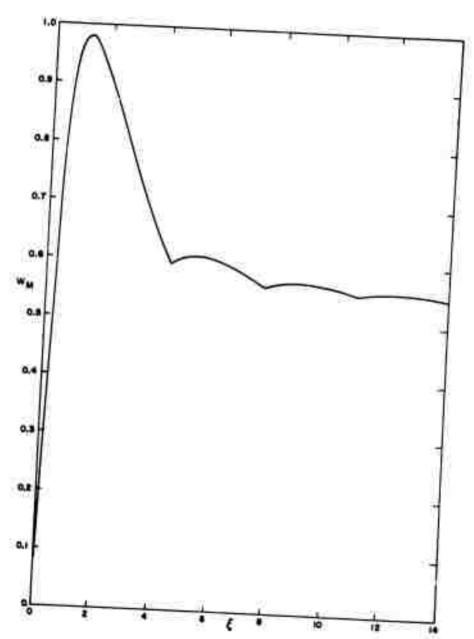


Figure 12. Spectrum of t  $\exp(-\epsilon t)$ -type function driving an ODOF system.

# APPENDIX A. — EVALUATION OF COEFFICIENTS OF PARTIAL FRACTION EXPANSION

If in (33) we multiply by s and let  $s \rightarrow \infty$ ,

$$A + C = 0 \tag{A-1}$$

If we let  $s \rightarrow 0$ ,

$$B/f^{2} + C + D + E = 1/f^{2}$$
 (A-2)

If we multiply (33) by  $(s+1)^3$  and let  $s \rightarrow -1$ ,

$$E = 1/(1+\xi^{\circ})$$
 (A-3)

If we multiply by s' + f' and let s - if,

$$Ai = + B = 1/(1 + i\xi)^{3}$$
 (A-4)

Equation (A-4) yields

$$(Aif + B)(1 + 3if - 3f^2 - ef^3) \equiv 1$$
 (A-5)

which upon expansion and equating real and imaginary parts yields the pair of equations

$$(1-3\frac{\pi^2}{2})B - \xi^2(3-\xi^2)A = 1$$
 (A-6)

and

$$(3 - \frac{\pi^2}{2})B + (1 - 3\xi^2)A = 0$$
 (A-7)

The determinant of the coefficient matrix reduces to  $(1 + s^2)^3$ . The solutions are readily found to be those given in (34a) and (34b). Finally D is obtained from (A-2)

### APPENDIX B --- INTEGRATION BY PARTS OF THE CONVOLUTION INTEGRAL

One defines .

$$I = \int_{2\pi} v^2 e^{-V} \sin \xi(\tau - v) dv$$
 (B-1)

$$J = \int_{0}^{\pi} v^{2} e^{-v} \cos f(\tau - v) dv$$
 (B-2)

$$K = \int_{0}^{\infty} ve^{-v} \sin \xi(\tau - v) dv$$
 (B-3)

$$L = \int_{X} ve^{-v} \cos^{\sigma}(\tau - v) dv$$
 (B-4)

$$M = \int_{0}^{T} e^{-V} \sin \xi(\tau - v) dv$$
 (B-5)

$$N = \int_{0}^{\pi} e^{-v} \cos F(\tau - v) dv$$
 (B-6)

where  $w = f I_*$  Integration by parts, retaining  $e^{-v}$  under the differential each time, yields the relations

$$I = 2K - \varepsilon J \tag{B-7}$$

$$J = -x^{2}e^{-\tau} + 2L + \xi I$$
 (B-8)

$$K = M - FL \tag{B-9}$$

$$L = -\tau e^{-\tau} + N + \xi K$$
 (B-10)

$$M = \sin \frac{\pi}{4} - \frac{\pi}{4}N \tag{B-11}$$

$$N = -e^{-\tau} + \cos \xi + \xi M \tag{B-12}$$

whose solution for I may be given as

$$I(1+\xi^{\circ})^{\circ} = \xi(1+\xi^{\circ})^{\circ} \cdot e^{-\tau} + 4\xi(1+\xi^{\circ})^{\circ} \cdot e^{-\tau} + 2\xi(3-\xi^{\circ})e^{-\tau} + \xi^{\circ}$$

$$2(1-3f^{\circ})\sin^{2}(-2f(3-f^{\circ})\cos^{2}(-(B-13))$$

### APPENDIX C. -- SCALE FACTORS OF ANALOG CIRCUIT

Table CI summarizes the essential, approximate scaling data obtained by trial using the circuit in figure 4. The potentiometer numbers in the five columns at the right are, of course, arbitrary. The circled numbers denote gains on the EAI 131-R analog computer. The values and gains are easily translated into corresponding values for computers having different gains available.

Table CI. Scale Factors and Potentiometer Settings for Analog Circuit

ξ	k	СW	Sı	42	44	46	48	52
0.2	2	500	2500	.5437 10	.1259 10	.5413 2	1.4	.4
0.5	1	125	250	.2718 10	.6796 2	.3383	5	.5
1.0	1	100	100	.2718 10	.6796 🔁	.5413	1.5	1 1
√3	1	100	100	.2718 10	.6796 2	.8120 2	$\Box$	.6 3
2.0	1	100	100	.2718 10	.6796 2	. 4331 (5)	<u>P</u>	8 5
3.0	1	100	100	.2718 10	.6796 2	.9744 5	( <b>i</b> )	.8 5
4.0	1	100	100	. 2718 10	.6796 2	.8661 10	J.	
5.0	0.2	125	100	.5437	.2718	. 5413 (5)	.25	.8 (5)
10.0	0.1	125	100	. 2718	.1359	.5413 10	.125	.8 20 .8 5 .8 10

#### APPENDIX D.—TABLE OF SYMBOLS

m mass

k spring constant (on analog diagram, used as time scale

factor)

x displacement

t time

F(t) forcing function

Fo maximum value of half-sine pulse forcing function

(excluding impulses)

β given decay constant of one forcing function

 $w_N = \sqrt{\frac{k}{m}}$  natural frequency of system

 $x_s = \frac{F_o}{k}$  static displacement

 $u = \frac{x}{x_g}$  response factor

tc duration of half-sine pulse

 $=\frac{t}{t_C}$  normalized time

 $\Psi_{\mathbf{F}} = \frac{\pi}{t_{\mathbf{C}}}$  angular frequency characteristic of forcing function

 $\gamma = \frac{N}{F}$  normalized natural frequency

TM time at which u reaches a global maximum

uM value of u at its global maximum
 F: strength of Delta impulse function

 $R = \frac{F_1}{F_2}$  ratio of impulse strength to half-sine pulse peak

 $w = \frac{4}{e^2} u$  scaled response factor

f = γπ scaled normalized natural frequency

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